

Comparison of Time Domain and Statistical IBIS-AMI Analyses

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9 Combinations of TX and RX Model Types

- AMI file has:
 - GetWave_Exists = True **Best for bit-by-bit simulation**
 - Init_Returns_Impulse = True **Best for statistical analysis**
 - 3 types: "Init-only", "GetWave-only", "Dual"
 - 3 TX * 3 RX = 9 combinations

Case #	TX			RX			Convolution Input	Statistical	Time Domain
	Getwave Exists	Init_Returns_Impulse	Meaning	Getwave Exists	Init_Returns_Impulse	Meaning			
1	FALSE	TRUE	Init-Only	FALSE	TRUE	Init-Only	3	OK	Static TX EQ, Static RX Eq
2	FALSE	TRUE	Init-Only	TRUE	FALSE	Getwave-Only	1 or 2	No RX EQ	Static TX EQ, Dynamic RX Eq
3	FALSE	TRUE	Init-Only	TRUE	TRUE	Dual	2	OK	Static TX EQ, Dynamic RX Eq
4	TRUE	FALSE	Getwave-Only	FALSE	TRUE	Init-Only	3	No TX EQ	Dynamic TX EQ, Static RX EQ
5	TRUE	FALSE	Getwave-Only	TRUE	FALSE	Getwave-Only	1,2,or 3	No TX or RX EQ	Dynamic TX EQ, Dynamic RX EQ
6	TRUE	FALSE	Getwave-Only	TRUE	TRUE	Dual	1	No TX EQ	Dynamic TX EQ, Dynamic RX EQ
7	TRUE	TRUE	Dual	FALSE	TRUE	Init-Only	iFFT(FFT(3)/FFT(2))	OK	Dynamic TX EQ, Static RX EQ
8	TRUE	TRUE	Dual	TRUE	FALSE	Getwave-Only	1	No RX EQ	Dynamic TX EQ, Dynamic RX EQ
9	TRUE	TRUE	Dual	TRUE	TRUE	Dual	1	OK	Dynamic TX EQ, Dynamic RX EQ

Simulation limitations

	Correct equalization of TX and RX modeled
	Correct equalization of TX and RX modeled: Assumes no adaptation in TX
	Assumes Static RX Equalization is a good representation of the RX: No adaptation
	Assumes Static RX EQ is a good representation of the RX: No Adaptation, Requires advanced math capabilities in Simulator
	Equalization data is missing

Case #	TX	RX	Statistical	Time Domain
1	Init Model Only	Init Model Only	OK	Static TX EQ, Static RX EQ
2	Init Model Only	Getwave Model Only	No RX EQ	Static TX EQ, Dynamic RX EQ
3	Init Model Only	Dual Model	OK	Static TX EQ, Dynamic RX EQ
4	Getwave Model Only	Init Model Only	No TX EQ	Dynamic TX EQ, Static RX EQ
5	Getwave Model Only	Getwave Model Only	No TX or RX EQ	Dynamic TX EQ, Dynamic RX EQ
6	Getwave Model Only	Dual Model	No TX EQ	Dynamic TX EQ, Dynamic RX EQ
7	Dual Model	Init Model Only	OK	Dynamic TX EQ, Static RX EQ
8	Dual Model	Getwave Model Only	No RX EQ	Dynamic TX EQ, Dynamic RX EQ
9	Dual Model	Dual Model	OK	Dynamic TX EQ, Dynamic RX EQ

Best Option

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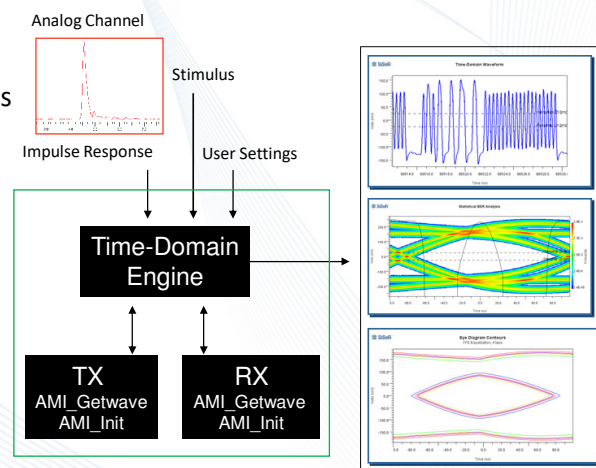
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Time-Domain Simulation

- **Inputs:**
 - Channel and buffer Impulse responses
 - User-defined input stimulus
 - Algorithmic models (**AMI_GetWave**)
- **Analysis Method:**
 - Waveform processing & convolution
- **Outputs:**
 - Bit pattern waveforms
 - Persistent eye diagrams
 - Eye height / width measurements
 - Eye contours @ probabilities
 - Equalized / unequalized responses



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Statistical Simulation

- **Inputs:**

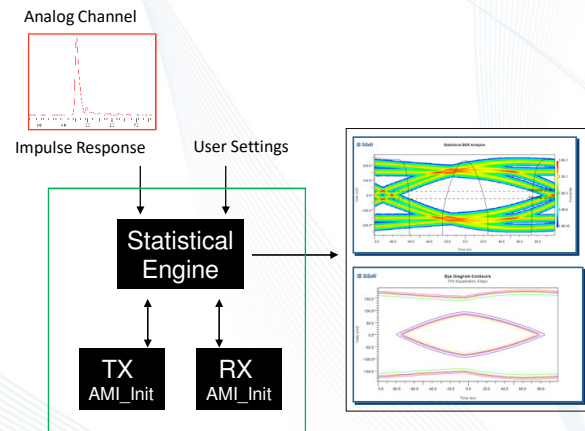
- Analog channel impulse response
- User selections for model parameters
- Algorithmic models (AMI_Init / impulse response processing)

- **Analysis Method:**

- Convolution engine (pulse response)

- **Outputs:**

- Statistical eye diagrams
- Eye height / width measurements
- Eye contours @ probabilities
- Equalized / unequalized responses



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Which IBIS-AMI Model Type is Best?

- **Need to evaluate suitability for modeling:**

- **Impairments:** The factors that harm the signal
 - Mostly in the channel
 - Statistical analysis has advantages
- **Corrective measures:** Signal improvements
 - Mostly inside the SerDes
 - Time domain has advantages

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Impairments To Be Modeled

Amplitude Impairment	Physical Cause
→ Inter-symbol interference (ISI)	Signal distortion (linear and nonlinear)
Crosstalk	Electromagnetic coupling in passive interconnect
Receiver sensitivity	Low signal amplitude causes decision latch to fail clock-data timing
Additive White Gaussian Noise (AWGN)	Shot noise in receiver amplifiers

Clock Impairment	Physical Cause
→ Random Jitter (RJ)	a. Shot noise in oscillator gain element b. Power supply noise modulating gate delays
Duty Cycle Distortion (clock) (DCD)	For half rate clock, duration difference between positive and negative half cycles
Duty Cycle Distortion (data)	Difference between data rise and fall times
Sinusoidal Jitter (SJ)	Clock noise on power supply modulating gate delays

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Corrective Measures To Be Modeled

Corrective Measure	Time Domain Behavior
→ TX FFE	May adapt in time domain, but this is rare
RX CTLE	Linear, time-invariant (LTI)
RX AGC	Adapts in time domain
RX Saturation	Not adaptive, but not time-invariant either
→ RX DFE	Adapts in time domain
Others...	

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Inter-Symbol Interference (ISI) Impairments

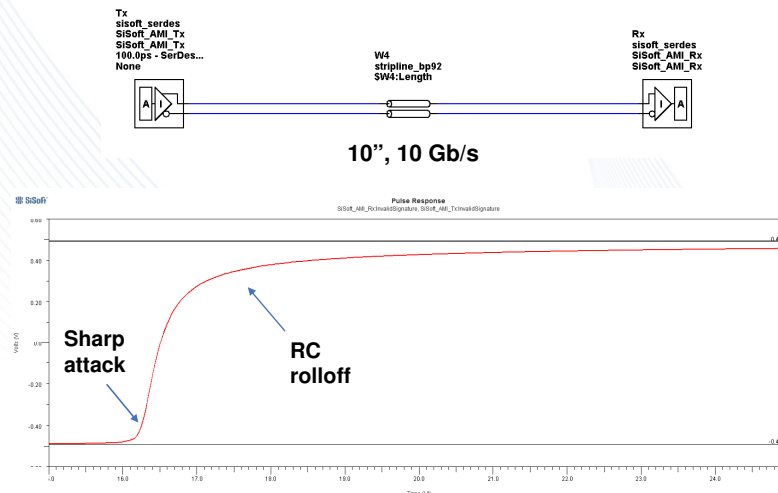
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Step Response Analysis



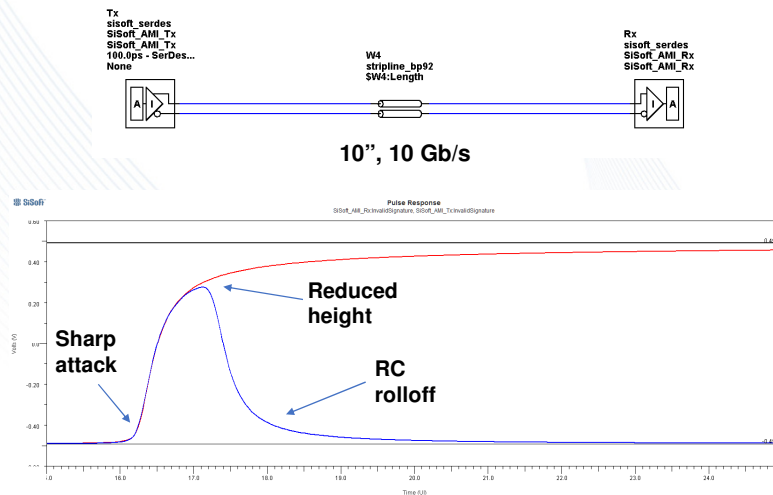
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Pulse vs. Step Responses



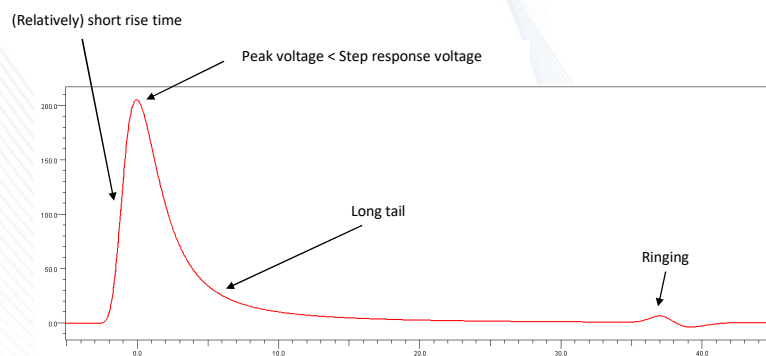
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Channel Pulse Response



- Requires accurate Tx/Rx analog models to correctly predict ringing impairment due to reflections

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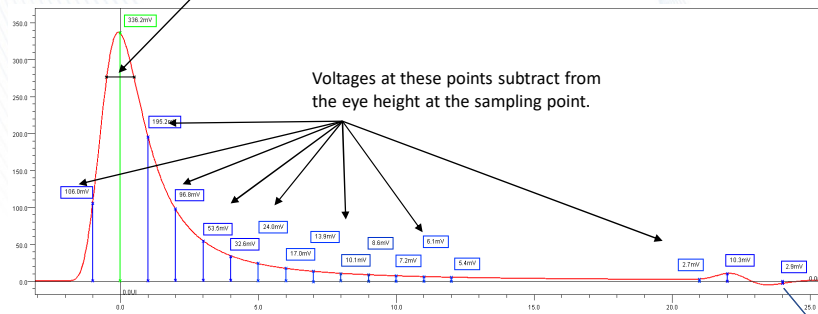
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Aligned Pulse Response and ISI

Hula hoop algorithm determines clock sampling time and main cursor height.
This is the maximum possible inner eye height.



$$\text{Inner Eye Height} = \text{main_cursor} - \sum |\text{ISI_voltages}|$$

- Voltage and time scales show ISI contributions
- Useful in evaluating EQ & predicting eye opening

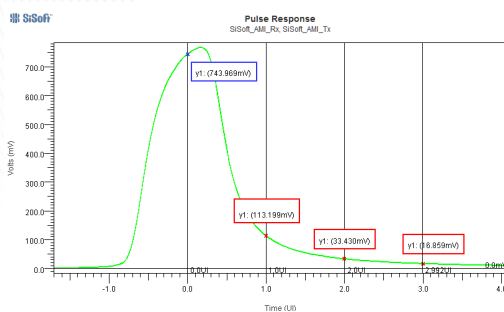
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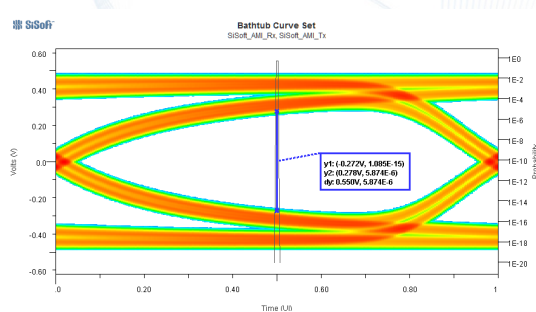
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Statistical ISI Inner Eye Quick Calculation



Prediction: 580mV

$$\text{Inner Eye Height} = \text{main_cursor} - \sum |\text{ISI_voltages}|$$



Simulated Actual: 550mV

A quick calculation gets us close, but small amounts of energy in the tail add up

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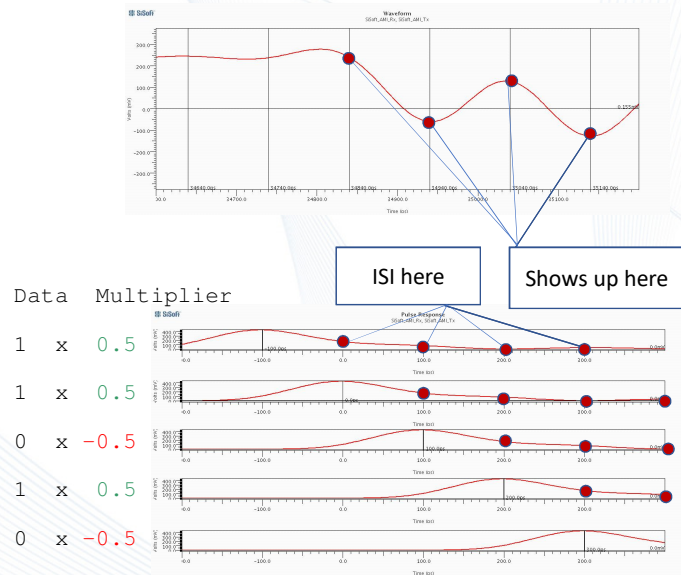
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Time Domain ISI

- Time domain waveform from impulse response
- Bit pattern modulated
- Linear superposition
- LTI assumed

Example bit pattern: 11010



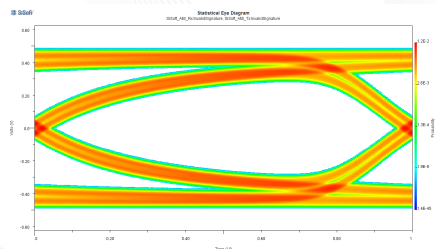
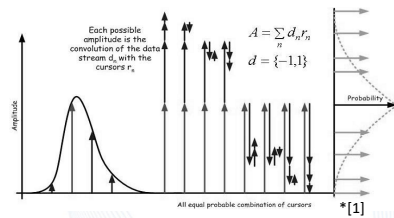
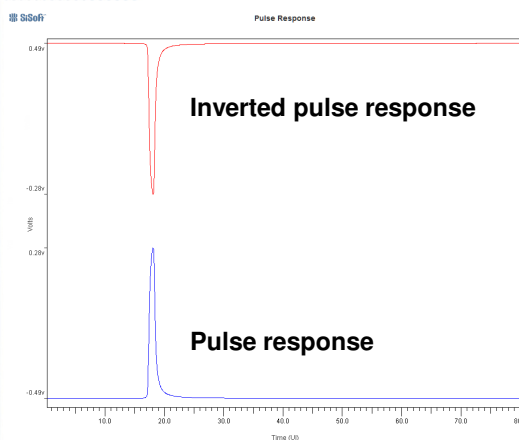
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Statistical ISI



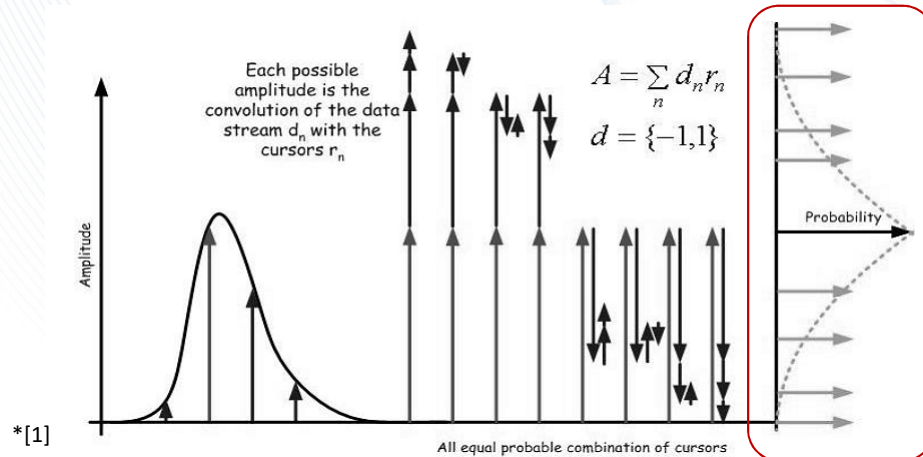
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All Possible LTI Combinations Evaluated



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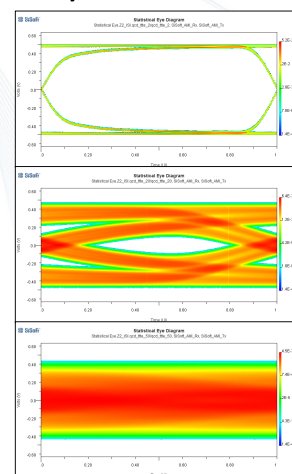
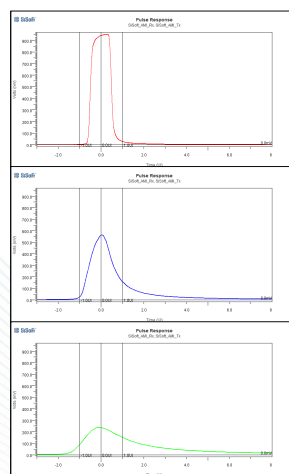
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Channels, Pulses and Statistical Eyes

**Short channel,
Minimal ISI**

**Medium channel,
Moderate ISI**

**Long channel,
Extreme ISI**



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Accounting for All ISI Scenarios

- A 28Gbps link may have a bit every 0.2 inches
- Many bits can be on the channel at once
- With reflections that number is multiplied
- Required impulse response may be many UI in length
- The bit pattern affects how these interact

To completely model all possible ISI scenarios we must try every possible bit pattern for the number of UI needed to capture all significant ISI

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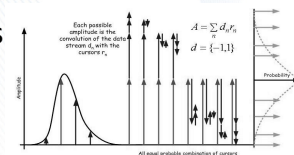
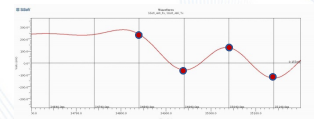
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Can We Account for All ISI Scenarios?

- Theoretically need to try 2^N patterns, where N is the number of UI before ISI becomes insignificant
 - Example: 24 UI NRZ impulse response must simulate $2^{24} = 16,777,216$ patterns, each 24 UI in length, total of 402,653,184 bit computations
- Time domain simulation
 - N-length patterns tested sequentially
 - PRBS helps reduce redundancies
 - Often able to simulate only a fraction of cursor combinations
- Statistical analysis
 - Able to directly calculate all 2^N cursor combinations
 - Efficient computation of channel response, not a circuit
 - May still have a practical upper limit for N



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Jitter and Noise Impairments

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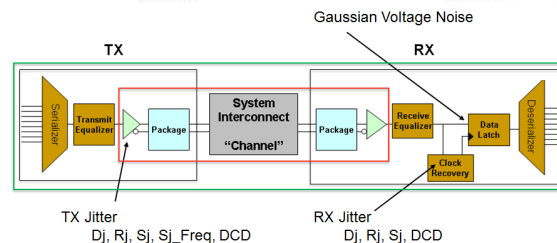
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Jitter and Noise in IBIS-AMI

- IBIS 6.1 provides multiple TX & RX impairments
- TX jitter directly modulates the TX output
 - Simulators jitter the stimulus pattern sent to the TX in time domain simulations
 - Statistical analysis convolves jitter with eye diagram
- RX jitter affects recovered clock behavior
 - Simulators combine jitter data with clock information returned by the RX
 - Statistical analysis convolves jitter with eye diagram
- RX noise affects sampling latch data input
- **Jitter and noise are handled by the simulator, not by the models**



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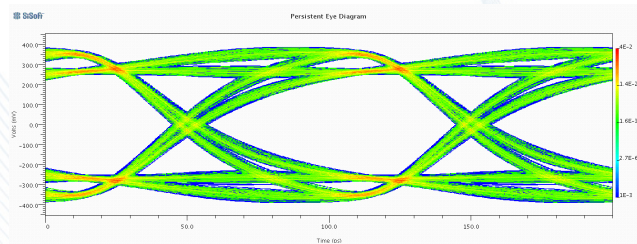
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Time Domain Eyes With and Without Tx Jitter

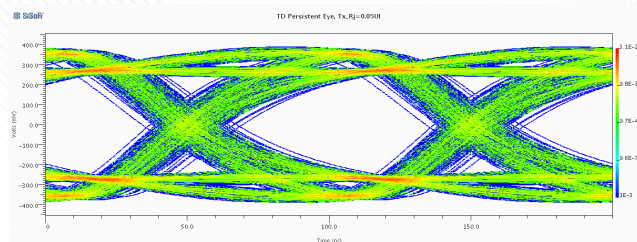
Random Jitter
(Tx_Rj) = 0



Only Impairment is
Inter-Symbol
Interference (ISI)

1e-3

Tx_Rj = 0.05UI



ISI + Jitter

1e-3

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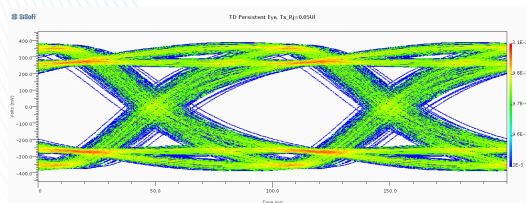
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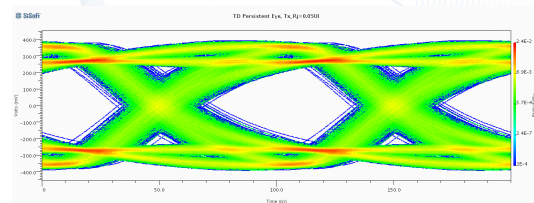
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Time Domain: How Many Bits to Simulate?

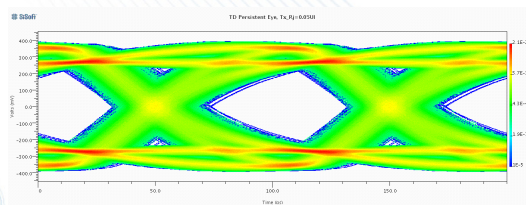
1,000 UI



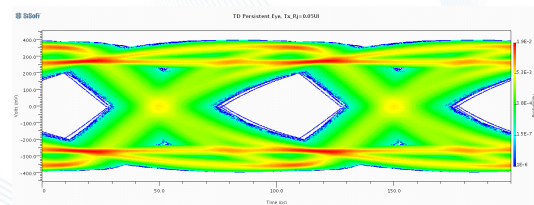
10,000 UI



100,000 UI



1,000,000 UI



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What Maximum BER Can We Tolerate?

- IEEE-802.3bj-KR4 FEC on $1e-5$
- IEEE-802.3bj-KR4 FEC off **$1e-12$** if low latency required
- OIF-CEI-56G FEC on $1e-4$
- OIF-CEI-56G FEC off **$1e-20$**
- PCIe-G3 **$1e-12$**
- PCIe-G4 **$1e-12$**
- DDR4 **$1e-12$** eye mask rules
- DDR5 TBD

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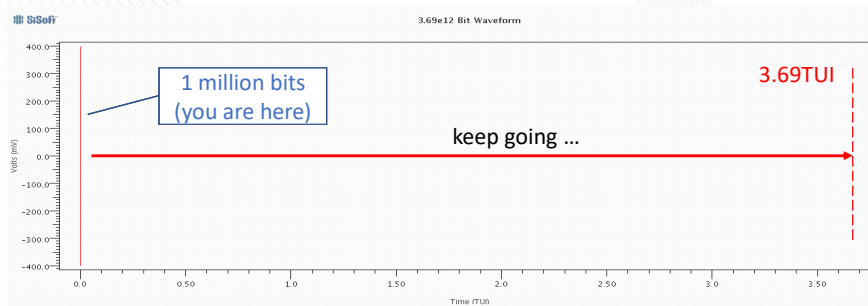
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How Many Error-Free Bits for $1e-12$ BER?

It's Not $1e12$

Confidence Level	90%	95%	99%
Maximum BER	$1e-12$	$1e-12$	$1e-12$
Error-free Bits Simulated *[2]	$3.00e12$	$3.69e12$	$5.30e12$

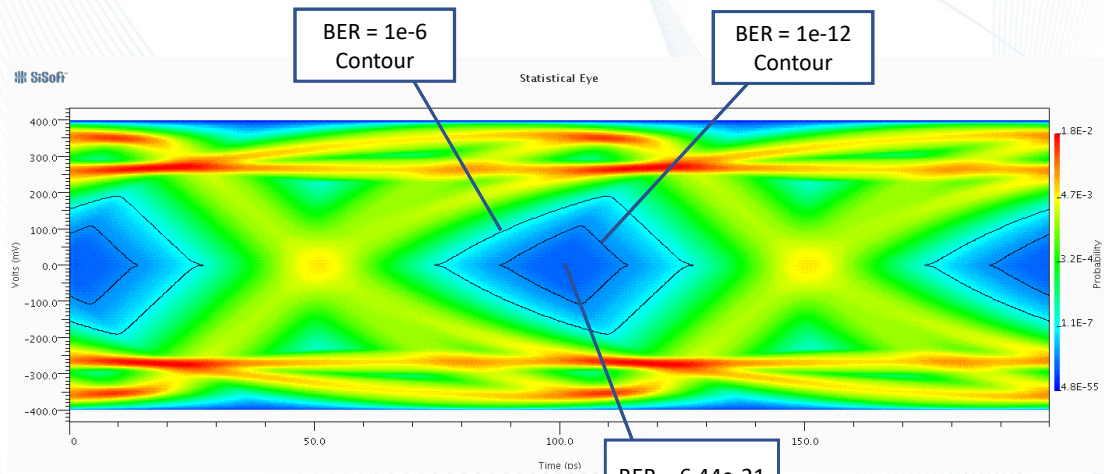


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Statistical Eye With ISI and Jitter



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Tx Corrective Measures

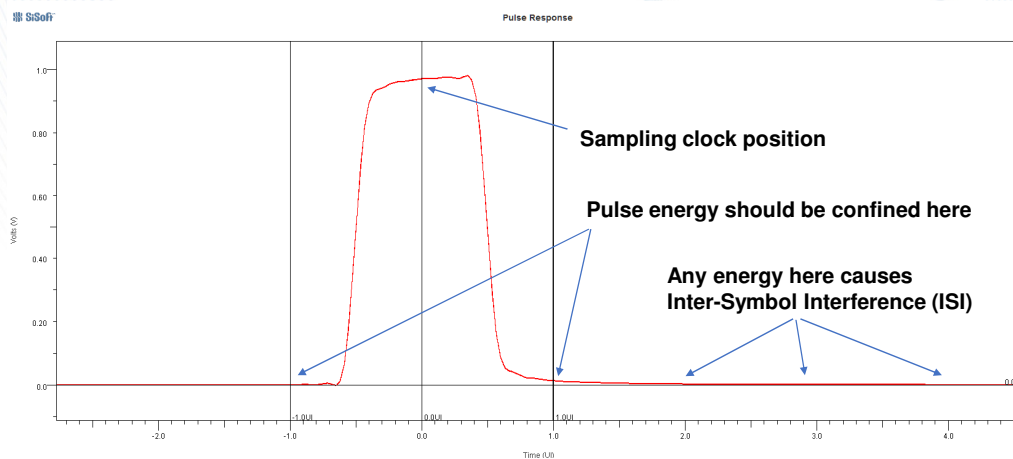
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Desired Pulse Response for Low ISI



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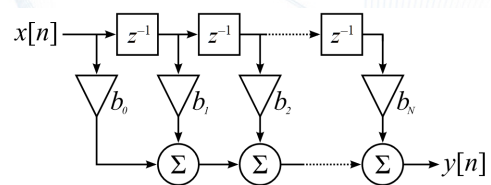
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Tx Feed-Forward Equalization (FFE)

- Usually implemented as taps spaced at the signal data rate
- Can precede the signal (pre-cursor), follow the signal (post-cursor), or both
- Typical configuration is 1 pre-cursor, 2 post-cursor taps



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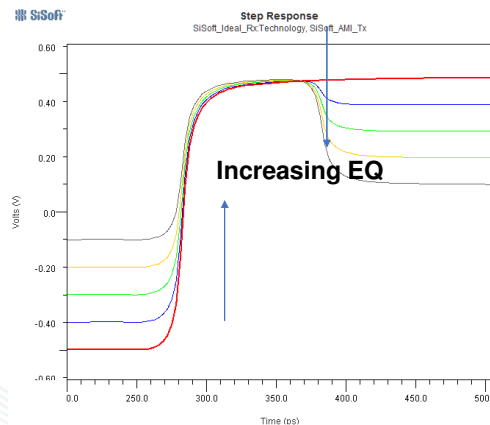
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TX FFE Equalization (1st post-cursor)

- Goal: boost high frequency content
- Transition occurs at full strength, then driver “pulls back” for subsequent bits
- TX EQ is often referred to as de-emphasis
- TX EQ always reduces the energy sent into the channel



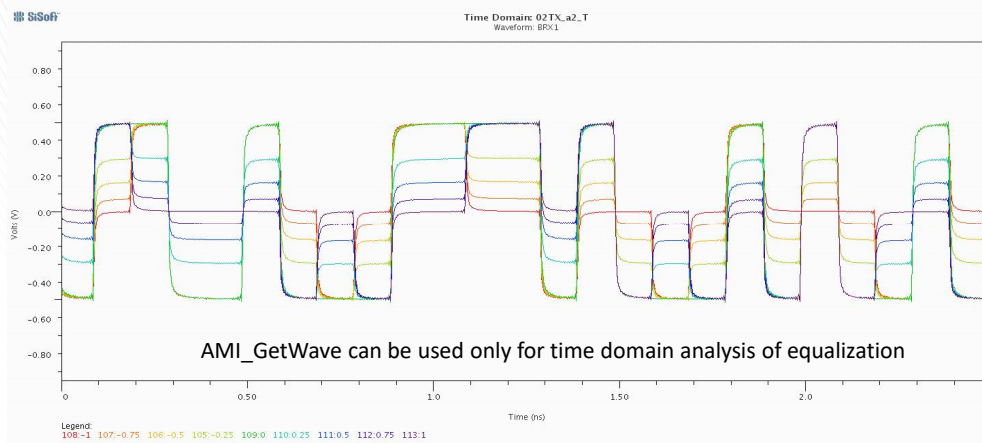
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AMI_GetWave Models Can Process Equalization Directly in Time Domain



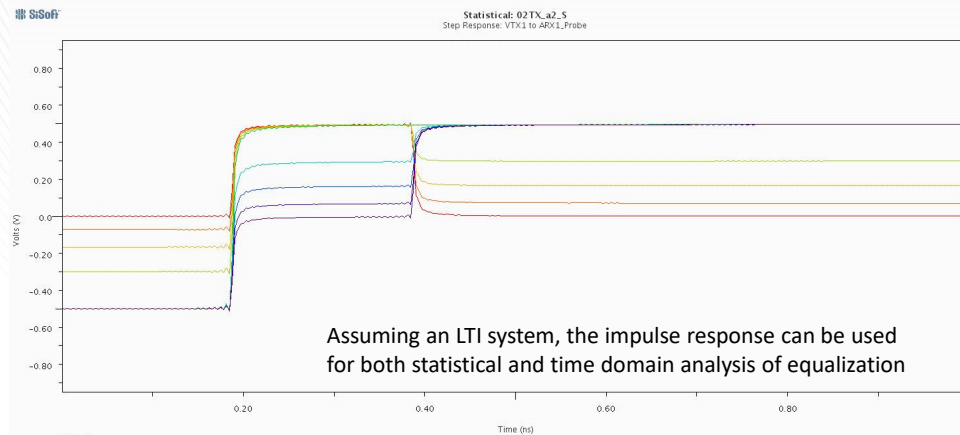
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AMI_Init Can Return Impulse Response for Equalization



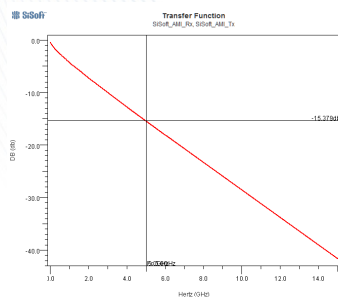
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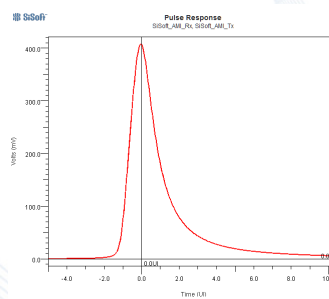
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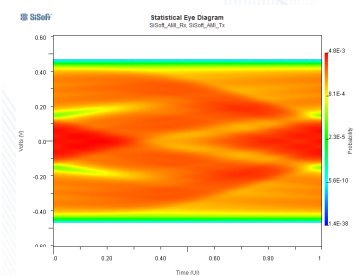
EQ Example: 20 inch channel, 10 Gb/s



15.3 dB loss



12+ bits of ISI



No EQ = No eye

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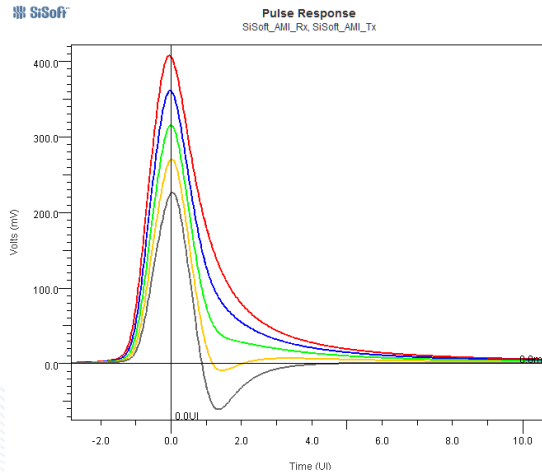
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Sweeping the 1st Post-cursor Pulse Response

Case	Cursor	1st Post
1	1.0	0.0
2	0.9	-0.1
3	0.8	-0.2
4	0.7	-0.3
5	0.6	-0.4

- Which case will give us the best eye?



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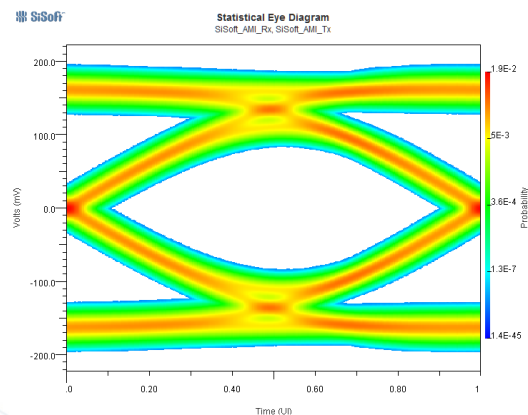
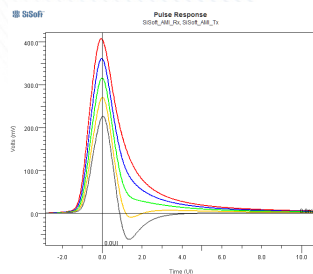
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Using Pulse Responses to Find TX Equalization

Row	Txtap_filter.0	Txtap_filter.1	Stat Eye Height (V)
1	1	0	0
2	.9	-.1	0
3	.8	-.2	0.0706985
4	.7	-.3	0.166147
5	.6	-.4	0.126204



Full Time Domain analysis not required

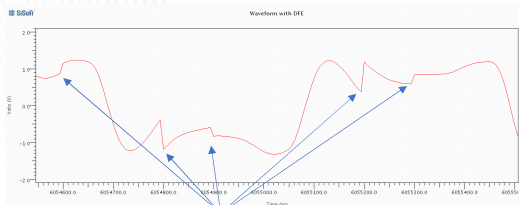
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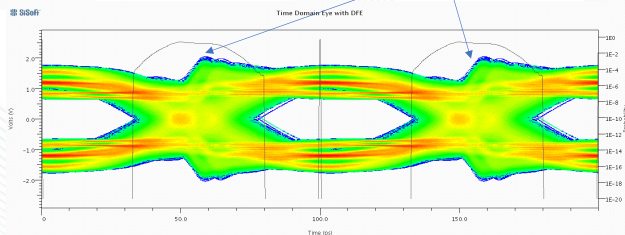
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AMI_GetWave Can Also Model Time-Variant Effects



RX Decision Feedback Equalizer (DFE) taps

- Adaptive corrections
 - DFE
 - CTLE
 - AGC
- Non-Linear Impairments
 - Saturation



RX DFE action visible in eye diagram

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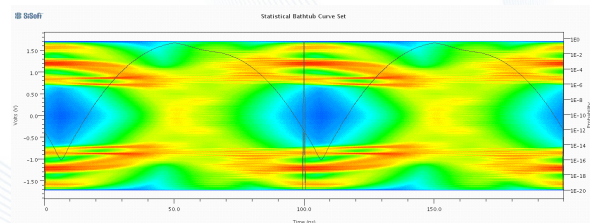
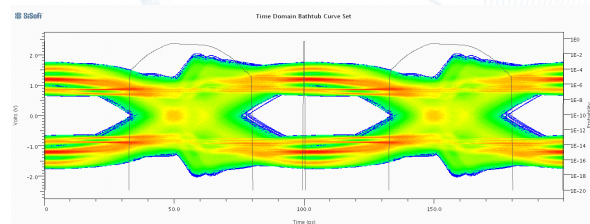
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Using Both Time Domain and Statistical Analysis

- No single analysis method models all impairments and all corrective measures well enough
- Many helpful techniques, eg.:
 - Statistical extrapolation of time domain
 - Get adapted settings from time domain and apply to statistical (can reduce Ignore_Bits)
 - Approximate adapted DFE in RX AMI_Init



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Conclusions

- IBIS-AMI time domain simulation with AMI_GetWave can precisely model non-linear effects such as DFE and saturation.
 - But it can be impossible to simulate enough bits in time domain to prove the low BER requirements of some technologies.
- IBIS-AMI statistical analysis can quickly evaluate very low BER.
 - But it can not precisely model time-variant effects such as DFE and saturation.
- It is good practice to use both analysis methods.

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Thank You

- Much content copied from:
 - *Pragmatic Signal Integrity Boot Camp*
 - Donald Telian, SiGuys
 - Michael Steinberger, SiSoft
 - Tripp Worrell, SiSoft
 - Todd Westerhoff, SiSoft
 - Graham Kus, SiSoft
 - Eric Brock, SiSoft
 - DesignCon 2017, Santa Clara, CA
- References
 - [1] Anthony Sanders, Mike Resso, John D'Ambrosia, *Channel Compliance Testing Utilizing Novel Statistical Eye Methodology*, DesignCon 2004
 - [2] Jeruchim, Michel C., Philip Balaban, and K. Sam Shanmugan, *Simulation of Communication Systems*, Second Edition, New York, Kluwer Academic/Plenum, 2000

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