

# Using IBIS in Non-ideal Power Supply Situations

- aka. Counter Proposal on BIRD 95

**Author: Lance Wang, Cadence**

**Contributors and Reviewers from:**

**Cadence, NCSU, Intel, Micron, ST Microelectronics**

**Version 1.0**

# Contributors and Reviewers

- Cadence
  - Shangli Wu, C. Kumar, Jilin Tan, Ken Willis, Dragoslav Milosevic
- North Carolina State University (NCSU)
  - Ambrish Varma, Paul Franzon
- Intel
  - Arpad Muranyi, Michael Mirmak
- Micron
  - Randy Wolff
- ST Microelectronics
  - Antonio Girardi

# Purpose

- IBIS needs to be used in Non-ideal power supply situations
- We need better solution than BIRD 95 since there is still a lot of concerns
  - BIRD 95 is adding VZDDQ and I/T table between Power and Ground. Link for BIRD 95:
    - <http://www.vhdl.org/pub/ibis/birds/bird95.6.txt>

# Outline

- Scientific Approach of New Proposal
  - BIRD 95 is inside
- Static Power Supply Change Test Results
- Notes
- Next Step

# General IBIS equation

General IBIS output current equation

$$-I_{\text{out}}(t) = K_u(t)I_{\text{pu}}(V_{\text{pu}}) + K_d(t)I_{\text{pd}}(V_{\text{pd}}) + I_{\text{pc}}(V_{\text{pc}}) + I_{\text{gc}}(V_{\text{gc}}) \quad (1)$$

Where:

$I_{\text{pu}}(V_{\text{pu}})$  is the Pullup Current @ Voltage between buffer Pullup Reference and Output (die point);

$I_{\text{pd}}(V_{\text{pd}})$  is the Pulldown Current @ Voltage between buffer Pulldown Reference and Output (die point);

$I_{\text{pc}}(V_{\text{pc}})$  is the Powerclamp Current @ Voltage between Power Clamp Reference and Output (die point);

$I_{\text{gc}}(V_{\text{gc}})$  is the Groundclamp Current @ Voltage between Ground Clamp reference and Output (die point);

## **The Development of Analog SPICE Behavioral Model Based on IBIS Model**

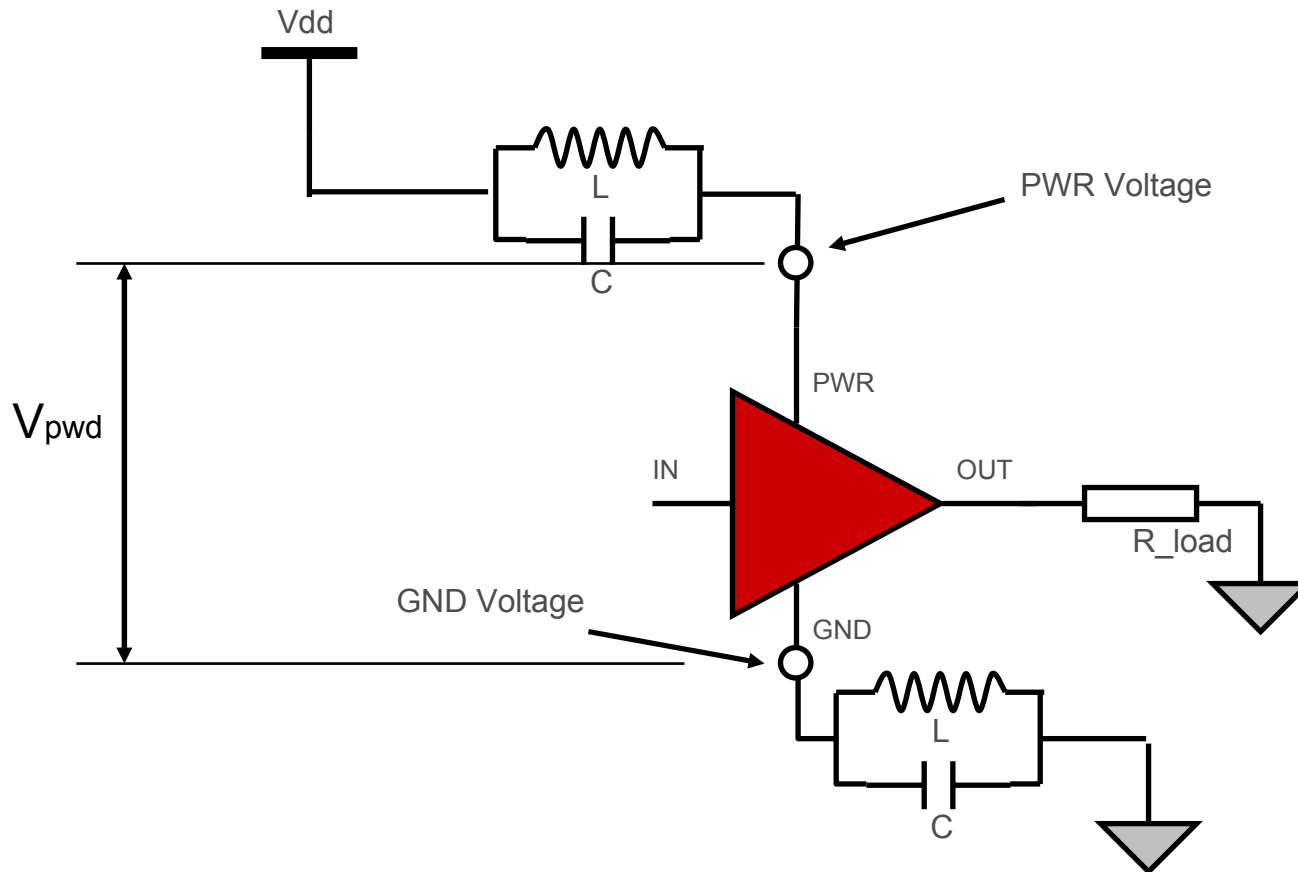
Based on article:

Ying Wang    Han Ngee Tan  
*School of EEE, Nanyang Technological University, Singapore 639798*  
*E-mail: p143278083@ntu.edu.sg*

<http://ieeexplore.ieee.org/iel4/6127/16377/00757386.pdf?arnumber=757386>

# Limitation of general IBIS equation

- Can not handle power/ground bounce for SSN
- Can not handle gate-modulation



# Equations for new proposal

Add power/ground voltage dependences to original IBIS equation

$$\begin{aligned} -I_{\text{out}}(t) = & K_u(t)I_{\text{pu}}(V_{\text{pu}}, V_{\text{pwdp}}) + K_d(t)I_{\text{pd}}(V_{\text{pd}}, V_{\text{pwdp}}) \\ & + I_{\text{pc}}(V_{\text{pc}}, V_{\text{pwdc}}) + I_{\text{gc}}(V_{\text{gc}}, V_{\text{pwdc}}) \end{aligned} \quad (2)$$

Where:

$I_{\text{pu}}(V_{\text{pu}}, V_{\text{pwdp}})$  is the Pullup Current @ Voltage between buffer Pullup Reference and Output (die point) and voltage between Pullup and Pulldown references ;

$I_{\text{pd}}(V_{\text{pd}}, V_{\text{pwdp}})$  is the Pulldown Current @ Voltage between buffer Pulldown Reference and Output (die point) and voltage between Pullup and Pulldown references ;

$I_{\text{pc}}(V_{\text{pc}}, V_{\text{pwdc}})$  is the Powerclamp Current @ Voltage between Power Clamp Reference and Output (die point) and voltage between Power and Ground Clamp references ;

$I_{\text{gc}}(V_{\text{gc}}, V_{\text{pwdc}})$  is the Groundclamp Current @ Voltage between Ground Clamp reference and Output (die point) and voltage between Power and Ground Clamp references ;

$$V_{\text{pu}} = V_{\text{cc}} - V_{\text{out}}, V_{\text{pd}} = V_{\text{out}} - V_{\text{cc}}, V_{\text{pc}} = V_{\text{pc}} - V_{\text{out}}, V_{\text{gc}} = V_{\text{out}} - V_{\text{gc}}.$$

# Equations for new proposal (cont.)

Assume power/ground bounces are small enough to use first order Taylor expansion

$$I_{pu}(V_{pu}, V_{pwp}) \approx I_{pu}(V_{pu}, V_{pwp0}) + \left. \frac{\partial I_{pu}}{\partial V_{pwp}} \right|_{V_{pwp0}} * (V_{pwp} - V_{pwp0})$$

$$I_{pd}(V_{pd}, V_{pwp}) \approx I_{pd}(V_{pd}, V_{pwp0}) + \left. \frac{\partial I_{pd}}{\partial V_{pwp}} \right|_{V_{pwp0}} * (V_{pwp} - V_{pwp0})$$

$$I_{pc}(V_{pc}, V_{pwc}) \approx I_{pc}(V_{pc}, V_{pwc0}) + \left. \frac{\partial I_{pc}}{\partial V_{pwc}} \right|_{V_{pwc0}} * (V_{pwc} - V_{pwc0})$$

$$I_{gc}(V_{gc}, V_{pwc}) \approx I_{gc}(V_{gc}, V_{pwc0}) + \left. \frac{\partial I_{gc}}{\partial V_{pwc}} \right|_{V_{pwc0}} * (V_{pwc} - V_{pwc0})$$

Where  $V_{pwp0}$  is Ideal Pullup/Pulldown Power Supply  
 $V_{pwc0}$  is Ideal Clamp Power Supply



# Equations for new proposal (cont.)

Substitute into equation (2)

$$\begin{aligned} -I_{out}(t) = & K_u(t)I_{pu}(V_{pu}, V_{pwp0}) + K_d(t)I_{pd}(V_{pd}, V_{pwp0}) + I_{pc}(V_{pc}, V_{pwc0}) + I_{gc}(V_{gc}, V_{pwc0}) \\ & + (K_u(t)Y_{pu} + K_d(t)Y_{pd})(V_{pwp}(t) - V_{pwp0}) + (Y_{pc} + Y_{gc})(V_{pwc}(t) - V_{pwc0}) \end{aligned} \quad (3)$$

where

$$Y_{pu} = \left. \frac{\partial I_{pu}}{\partial V_{pwp}} \right|_{V_{pwp0}}$$

$$Y_{pd} = \left. \frac{\partial I_{pd}}{\partial V_{pwp}} \right|_{V_{pwp0}}$$

$$Y_{pc} = \left. \frac{\partial I_{pc}}{\partial V_{pwc}} \right|_{V_{pwc0}}$$

$$Y_{gc} = \left. \frac{\partial I_{gc}}{\partial V_{pwc}} \right|_{V_{pwc0}}$$

# Equations for new proposal (cont.)

Re-arrange equation (3) to

$$-I_{out}(t) = -I_{out}(t, V_{out})|_{V_{pwp0}, V_{pwc0}} + (K_u(t)Y_{pu} + K_d(t)Y_{pd}) * V_{pwp}(t) + (Y_{pc} + Y_{gc}) * V_{pwc}(t) - I_{offset}(t) \quad (4)$$

where

$$\begin{aligned} -I_{out}(t, V_{out})|_{V_{pwp0}, V_{pwc0}} &= K_u(t)I_{pu}(V_{pu}, V_{pwp0}) + K_d(t)I_{pd}(V_{pd}, V_{pwp0}) \\ &+ I_{pc}(V_{pc}, V_{pwc0}) + I_{gc}(V_{gc}, V_{pwc0}) \\ I_{offset}(t) &= (K_u(t) * Y_{pu} + K_d(t) * Y_{pd}) * V_{pwp0} + (Y_{pc} + Y_{gc}) * V_{pwc0} \end{aligned}$$

# Equations for new proposal (cont.)

Equation (4) could easily include gate modulation effect

$$-I_{out}(t) = -I_{out}(t, V_{out})|_{V_{pwp0}, V_{pwc0}} + (K'_u(t)Y_{pu} + K'_d(t)Y_{pd}) * V_{pwp}(t) + (Y_{pc} + Y_{gc}) * V_{pwc}(t) - I'_{offset}(t) \quad (6)$$

where

$$-I_{out}(t, V_{out})|_{V_{pwp0}, V_{pwc0}} = K'_u(t)I_{pu}(V_{pu}, V_{pwp0}) + K'_d(t)I_{pd}(V_{pu}, V_{pwp0})$$

$$+ I_{pc}(V_{pc}, V_{pwc0}) + I_{gc}(V_{gc}, V_{pwc0})$$

$$I'_{offset}(t) = (K'_u(t) * Y_{pu} + K'_d(t) * Y_{pd}) * V_{pwp0} + (Y_{pc} + Y_{gc}) * V_{pwc0}$$

$$K'_u(t) = K_u(t) * K_{ssn-pullup}$$

$$K'_d(t) = K_d(t) * K_{ssn-pulldown}$$

Where Kssn-pullup/Kssn-pulldown are coefficients for Gate modulation

# Equations for new proposal (cont.)

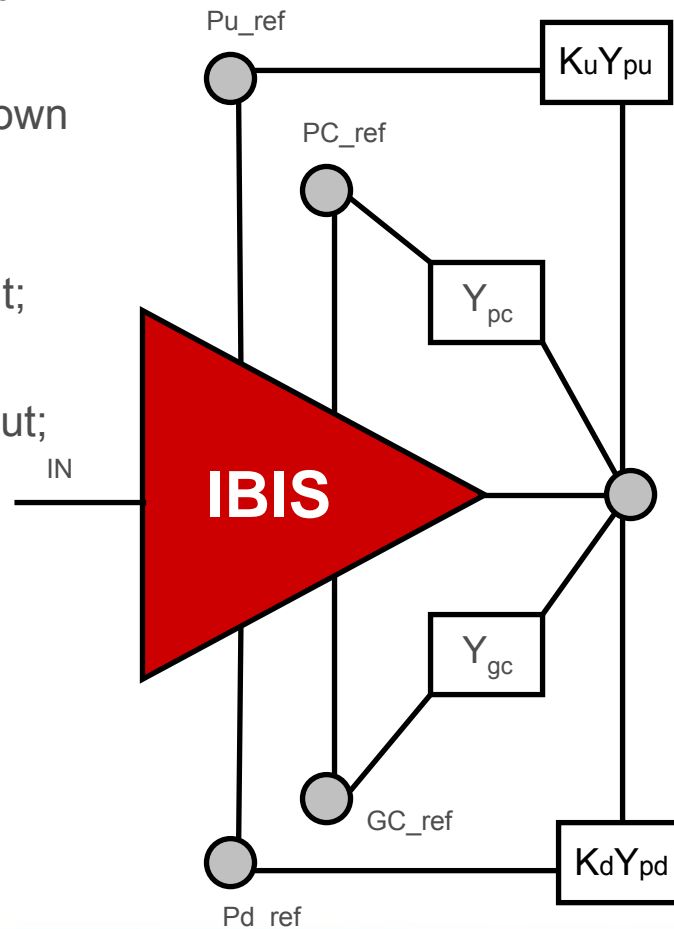
## Implementation

$Y_{pu}$  should be connected between Pullup Reference and Output;

$Y_{pd}$  should be connected between Pulldown Reference and Output;

$Y_{pc}$  should be connected between PowerClamp Reference and Output;

$Y_{gc}$  should be connected between GroundClamp Reference and Output;



# Equations for new proposal (cont.)

- Transconductance extraction

- $Y_{pu}$

- Average slope table of  $I_{out}/V_{pwpd}$  curve between  $V_{pwpd0} \pm \Delta V_p$

- $Y_{pd}$

- Average slope table of  $I_{out}/V_{pwpd}$  curve  $V_{pwpd0} \pm \Delta V_p$

- $Y_{pc}$

- Average slope table of  $I_{pc}/V_{pwc}$  curve between  $V_{pwc0} \pm \Delta V_c$

- $Y_{gc}$

- Average slope table of  $I_{gc}/V_{pwc}$  curve between  $V_{pwc0} \pm \Delta V_c$

Where  $\Delta V_p = (V_{pwpd} - V_{pwpd0}) * 0.05$ ,  $\Delta V_c = (V_{pwc} - V_{pwc0}) * 0.05$

# BIRD 95 is inside

Further simplification equation (4) to BIRD 95 approach

$$-I_{out}(t) \approx -I_{out}(t, V_{out})|_{V_{pwp0}, V_{pwc0}} + Y_{vddq} * V_{pwp}(t) - I'_{offset}(t) \quad (5)$$

where

$$\begin{aligned} -I_{out}(t, V_{out})|_{V_{pwp0}, V_{pwc0}} &= K_u(t)I_{pu}(V_{pu}, V_{pwp0}) + K_d(t)I_{pd}(V_{pd}, V_{pwp0}) \\ &+ I_{pc}(V_{pc}, V_{pwc0}) + I_{gc}(V_{gc}, V_{pwc0}) \\ I'_{offset}(t) &= (K_u(t) * Y_{pu}) * V_{pwp0} + Y_{pc} * V_{pwc0} \\ Y_{vddq} &= (K_u(t) * Y_{pu} + Y_{pc}) \end{aligned}$$

Assumptions for this simplification:

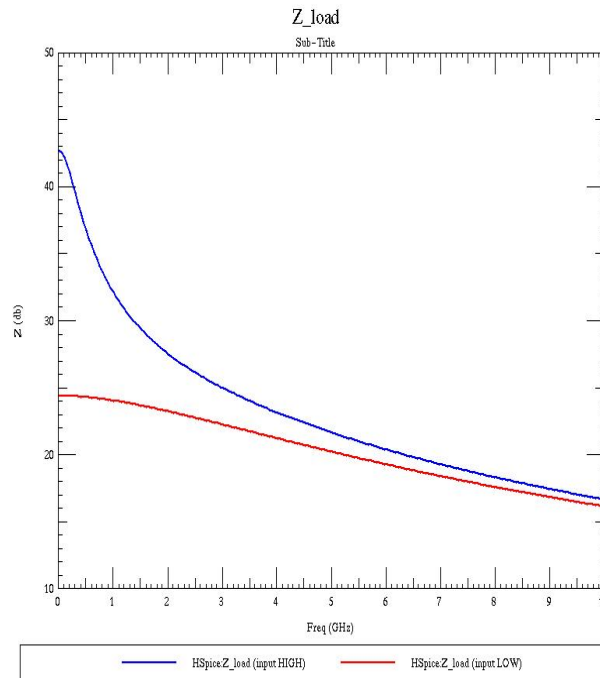
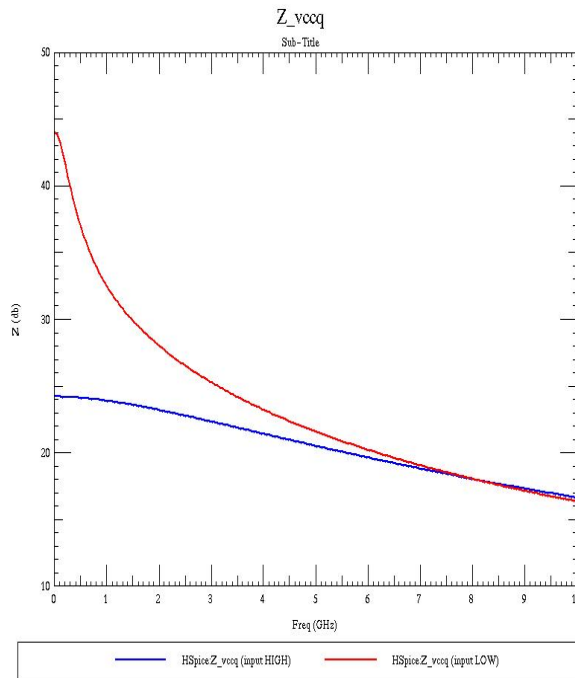
1. No ground bounce ( $V_{gnd}(t) = V_{gnd0} = 0$ )
2. All additional offset current go to ground instead of  $I_{out}$ . (This could be applied only for Pre-drive stage or Open terminations.)
3. Transconductance  $Y_{vddq}$  could be connected ONLY between power and ground if condition 2 is valid

# Equations for new proposal (cont.)

- Further more
  - We added  $Y$  based on DC portion of Power Supply Changes
  - We need to add  $Y_{AC}$  based on AC portion of Power Supply Changes
  - $Y_{AC}$  could be voltage controlled capacitance based on our current study result
  - Study is continued for this new proposal
  - More info links
    - <http://www.vhdl.org/pub/ibis/futures/bird95-cap.ppt>
    - [http://www.vhdl.org/pub/ibis/futures/ST\\_Vgs\\_Presentation.pdf](http://www.vhdl.org/pub/ibis/futures/ST_Vgs_Presentation.pdf)

# Equations for new proposal (cont.)

- Test Results of Dynamic Output Impedances
  - Input State Dependent Impedance measured between Power and Output ( $Z_{vccq}$ ), Ground and Output ( $Z_{load}$ )

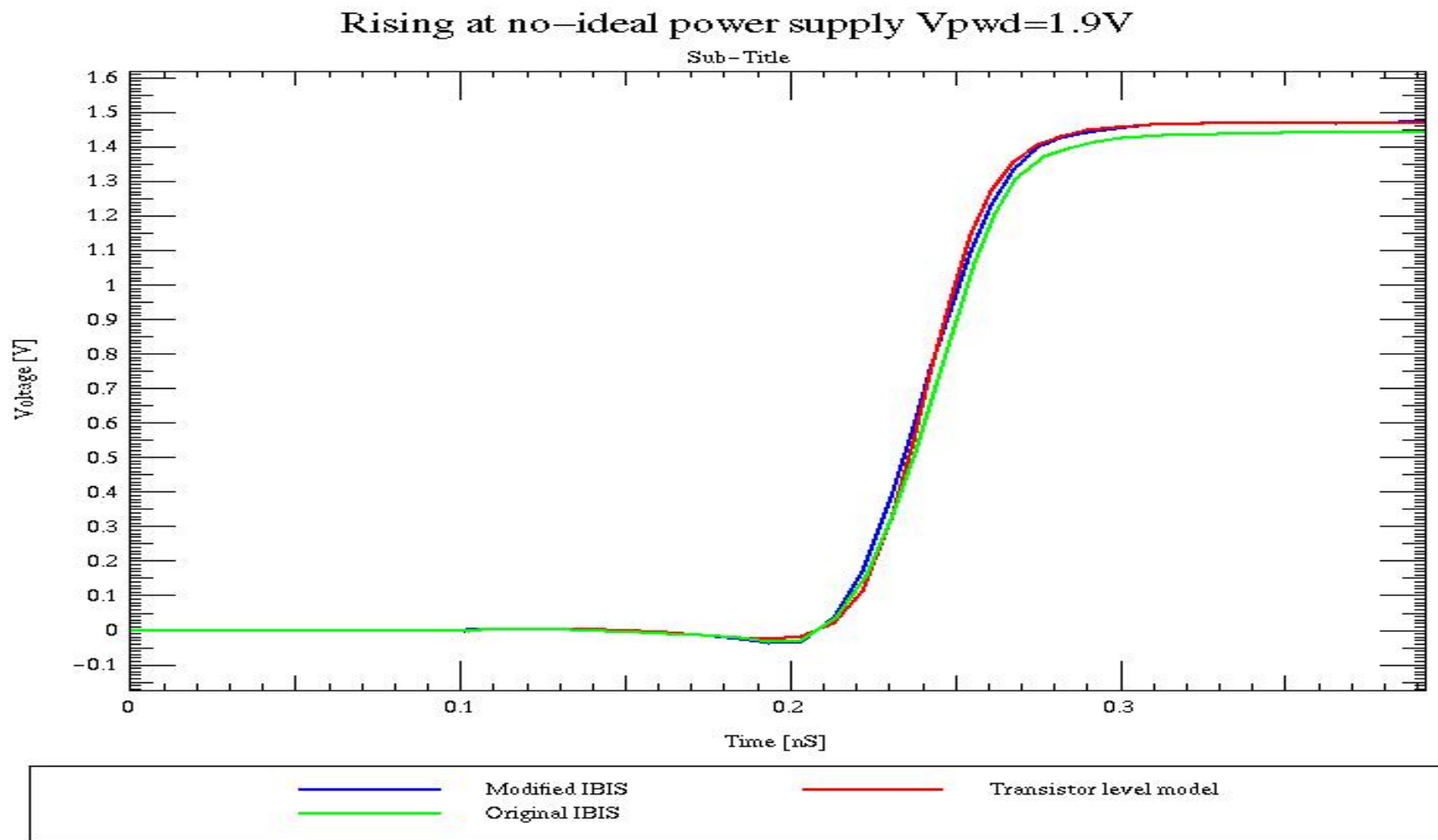




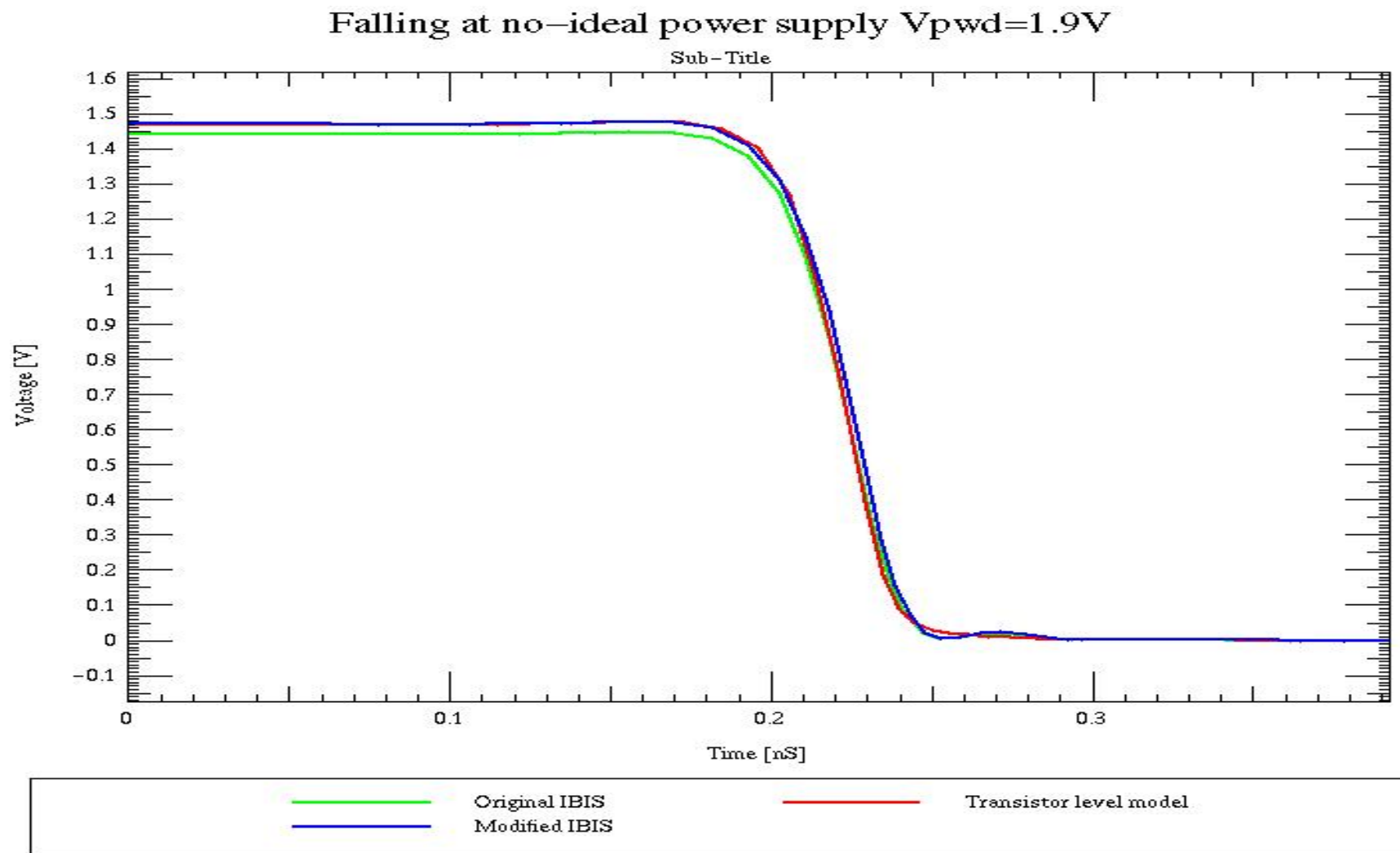
# Static Power Supply Change Test Results

- Case1: Rising and Falling @  $V_{\text{pwr}} = 1.9\text{V}$
- Case2: Rising and Falling @  $V_{\text{pwr}} = 1.7\text{V}$
- Using IBIS model with I/V and V/T table characterized @  $V_{\text{pwr}}=1.8\text{V}$

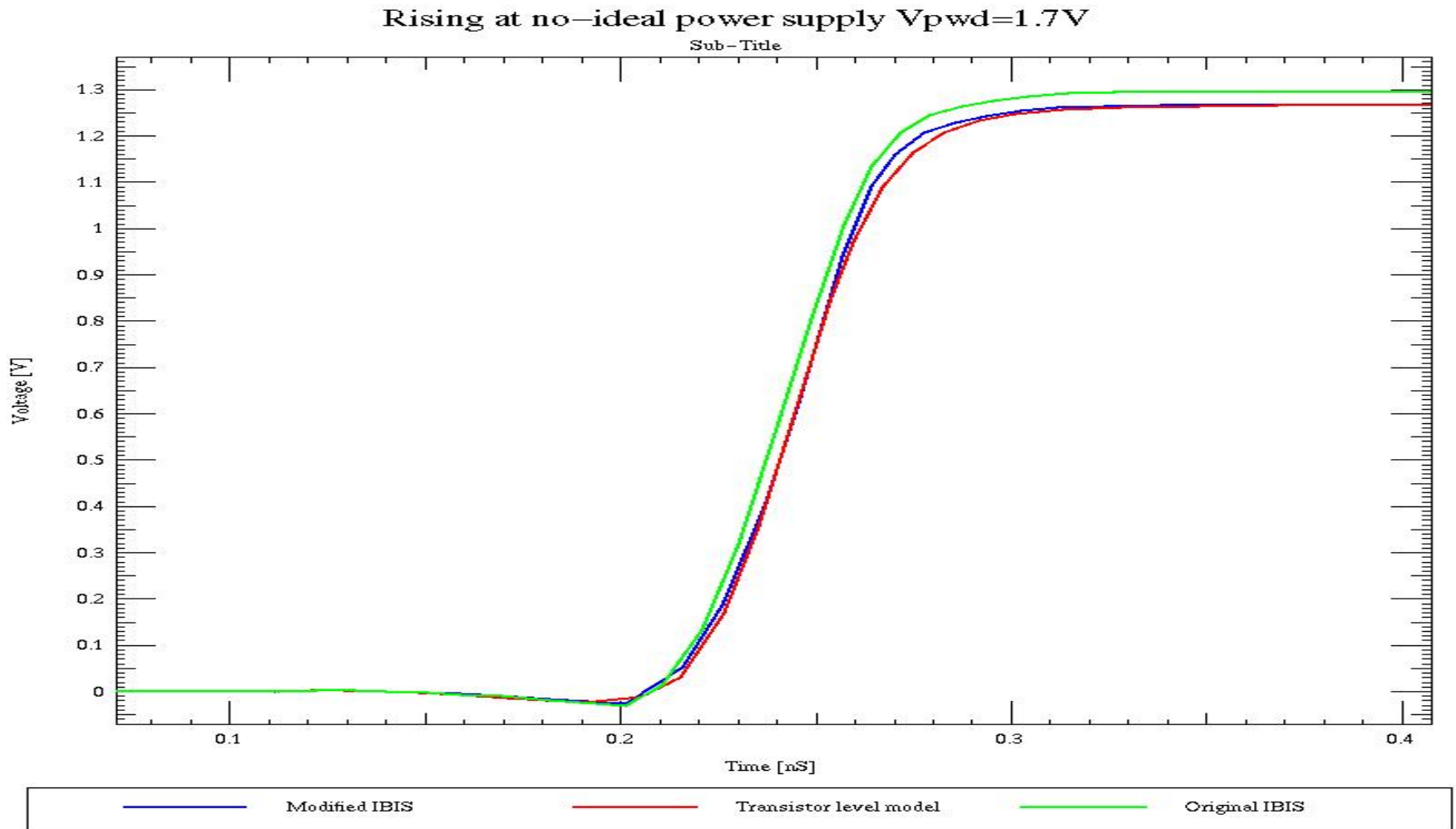
# Static Power Supply Changed to 1.9V



# Static Power Supply Changed to 1.9V

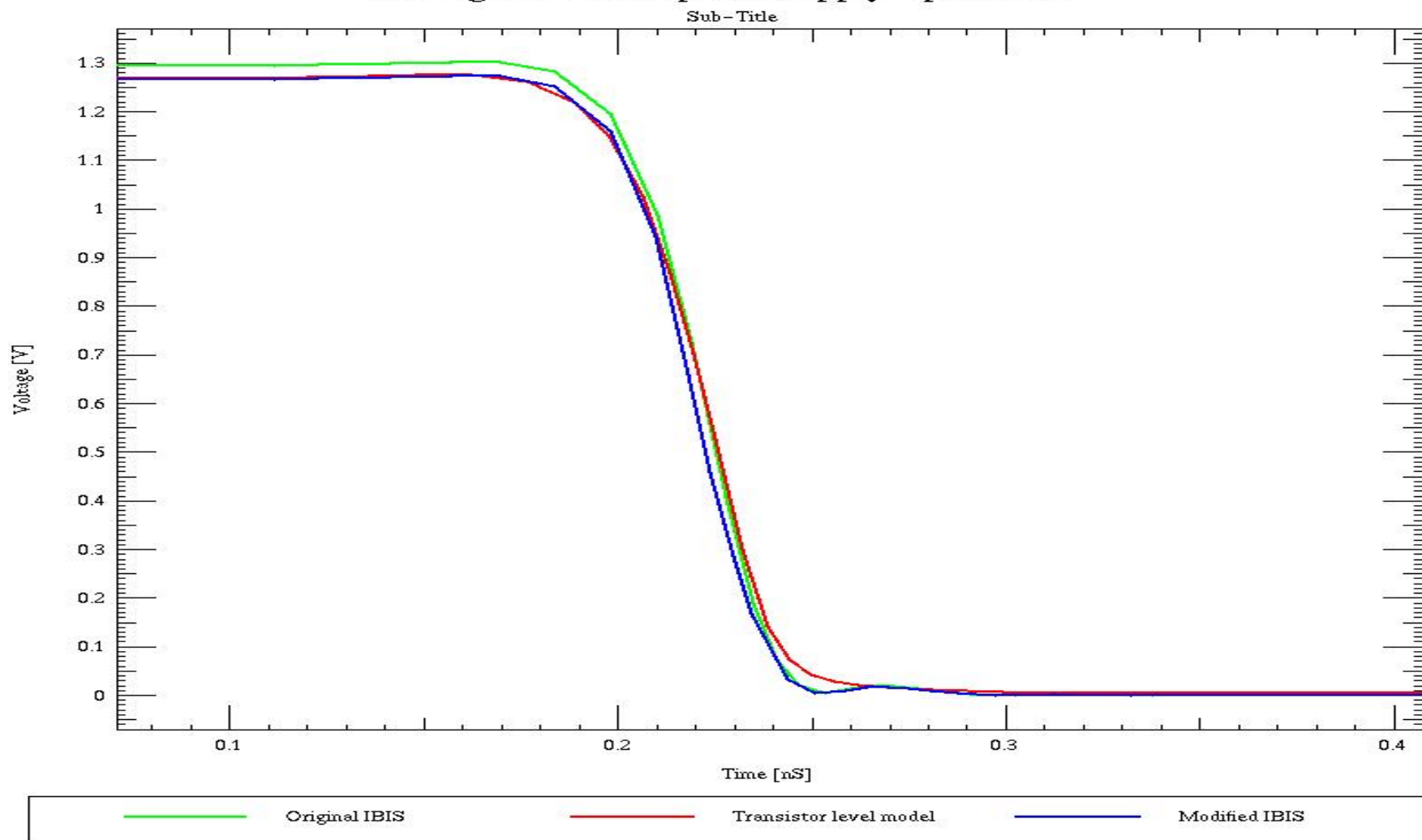


# Static Power Supply Changed to 1.7V



# Static Power Supply Changed to 1.7V

Falling at no-ideal power supply  $V_{pwr}=1.7V$



# Notes

- Theoretically, this proposal contains the goals both BIRD 98/97 and BIRD 95 try to achieve
  - Showed the limitations of BIRD 95
- Using IBIS in non-ideal power supply network could be solved in one proposal. It could be better to be solved as a big picture view

# Next Steps

- Continue to study and develop  $Y_{AC}$
- Test them
- Discuss it in IBIS committee
- Propose new BIRD or add-on to BIRD 98/97 when it is properly

**Welcome any comments and  
inputs for the technical  
discussions!**

**Thanks!**

**Thanks all efforts from the  
contributors and reviewers.**

**And**

**The contributions from all  
the members in IBIS are  
needed!**